

Features Used

[GRAPH], solve(), expand(), getDenom(), zeros(), NewProb, [Y=], getNum(), factor(), [], 3D graph, abs(), [WINDOW], cFactor(), NewData, cZeros(), real(), imag()

Setup

1, NewFold laplace

Laplace Analysis: The s-domain

This chapter demonstrates the utility of symbolic algebra by using the Laplace transform to solve a second-order circuit. The method requires that the circuit be converted from the time-domain to the s-domain and then solved for V(s). The voltage, v(t), of a sourceless, parallel, RLC circuit with initial conditions is found through the Laplace transform method. Then the solution, v(t), is graphed.

This chapter also shows how to find and plot the poles and zeros of a circuit's transfer function H(s) to gain insight to the frequency response.

Topic 27: RLC Circuit

Given the circuit shown in Figure 1, find v(t) for t>0 when v(0)=4 V and i(0)=1 A.



Figure 1. Simple parallel RLC circuit

Convert the components to their s-domain equivalents. Remember, the time-domain components map to their s-domain counterparts as shown in Figure 2.



Figure 2. Time-domain to s-domain mappings

Note that the initial voltage and current transform into equivalent sources in the s-domain. The circuit in the s-domain is shown in Figure 3.



Figure 3. s-domain equivalent of the circuit in Figure 1

Using Kirchhoff's current law to sum the currents out of the top node, the equation is

$$\frac{v}{4} + \frac{v}{3s} + \frac{v}{\frac{24}{s}} = \frac{4}{24} - \frac{1}{s} \rightarrow n1$$

- 1. Clear the TI-89 by pressing 2nd [F6] 2:NewProb ENTER.
- **2.** Enter the equation above (screen 1).

 $v \div 4 + v \div (3s) + v \div (24 \div s) = 4 \div 24 - 1$ $\div s \text{STOP} n1$

3. The s-domain voltage is found with **solve**(**n1**,**v**) as shown in screen 2.

CATALOG solve(n1 , v) STO> eqn

4. Enter **expand**(**eqn**) to put **eqn** in a form for easy calculation of the inverse Laplace transform via a table lookup (screen 3).

This must be an overdamped circuit since there are two real poles. The answer should contain two decaying exponents. From a Laplace transform table, the solution is

$$v(t) = 20e^{-4t} - 16e^{-2t} t \ge 0$$

This answer is in the expected mathematical form. How does v(t) appear as a function of time?



5. To obtain a graph, press ● [Y=] and enter the expression for v(t) as shown in screen 4. Note that x is substituted for t using the "with" operator, [], since the Y= Editor requires equations to be expressed as functions of x.

 $\mathbf{20 \times \bullet} [e^x] (-) \mathbf{4t}) [-] \mathbf{16 \times \bullet} [e^x] (-) \mathbf{2t}) [] \mathbf{t} = \mathbf{x}$

6. Now press F2 6:ZoomStd to see the graph (screen 5).

It appears to be a typical overdamped response!

- **7.** To zoom in for a closer look, press [WINDOW] and set the range of **x** to be 0 to 4 (screen 6).
- **8.** Now, press [GRAPH] to see the graph of v(t) as shown in screen 7.
- **9.** Press **F1 9:Format** and specify **ON** for **Grid** and **Labels** (screens 8 and 9).

Note: x is substituted for **t** using the **with** operator, [], since the **Y**= Editor requires equations to be expressed as functions of **x**.



Topic 28: Critical Damping

Given the circuit of Topic 27, change the value of the resistor so that the circuit will be critically damped.

Assign the resistor value as r (see Figure 4) and let the TI-89 compute the node voltage v(s) in terms of r.



Figure 4. Circuit of Figure 3 with the 4 Ω resistor changed to **r**

The nodal equation in the s-domain is

$$\frac{v}{r} + \frac{v}{3s} + \frac{v}{\frac{24}{s}} = \frac{4}{24} - \frac{1}{s} \rightarrow n1$$

1. Return to the Home screen, and enter this as shown in screen 10.

 $v \div r + v \div (3s) + v \div (24 \div s) = 4 \div 24 - 1$ $\Rightarrow s \land TOP n1$

- Solve for the node voltage with solve(n1,v) → eqn (screen 11).
- For critical damping, the time constants of the two exponentials of v(s) must be real and equal. To determine this condition, the two roots of the denominator of v(s) are found and set equal, and the resulting equation is solved for the required value of r. Get the denominator with getDenom() as shown in screen 12.

CATALOG getDenom(v | eqn) ST0► eqn2

4. Solve for values of **s** which are the roots of the denominator using the **zeros()** command as shown in screen 13.

CATALOG zeros(eqn2 , s) STO> z





5. Set the two roots equal to each other and solve for **r** as shown in screen 14.

CATALOG solve(z 2nd [[] 1 2nd []] = z 2nd [[] 2 2nd []] , r

Since negative resistances are not physically possible, the answer must be $\mathbf{r} = 3\sqrt{2}$.

6. To get the floating point approximation, press ● [≈] as shown in screen 15.

So $\mathbf{r} = 4.2$ will give critical damping.

 $\begin{array}{c} \hline F_{1}^{F_{1}}, F_{2}^{F_{2}}, e^{F_{3}}, e^{F$



Topic 29: Poles and Zeros in the Complex Plane

Given that

$$H(s) = \frac{s^4 + 14s^3 + 74s^2 + 200s + 400}{s^4 + 10s^3 + 49s^2 + 100s}$$

find and plot the poles and zeros.

1. Enter **h**(**s**) as shown in screen 16.

(s ∩ 4 + 14s ∩ 3 + 74s ∩ 2 + 200s + 400)) ÷ (s ∩ 4 + 10s ∩ 3 + 49s ∩ 2 + 100s)) STO▶ h

2. A quick way to see the poles and zeros is to factor **h**(**s**) as shown in screen 17.

CATALOG factor(h)

 However, since factor() doesn't give complex factors, use cFactor() to get more information about h(s) (screen 18).

CATALOG cFactor(h, s)

Press \bigodot () to see the rest of the terms of $h(\boldsymbol{s}).$ The complete answer is

$$\frac{(s-2(-3+i))(s-(-1+3i))(s+1+3i))(s+2(3+i))}{s(s+4)(s-(-3+4i))(s+3+4i)}$$

$$(16) \begin{bmatrix} F_{12} & F_{22} & F_{32} & F_{42} & F_{5} & F_{5} \\ \hline F_{10} & F_{12} & F_{22} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\ \hline F_{12} & F_{12} & F_{12} & F_{12} & F_{12} \\$$

Note: To enter *factor(*, press F2) *2:factor(*.

$$(17) \frac{ \begin{bmatrix} F_{2}^{+}, \end{bmatrix} \end{bmatrix} \\ 1 + F_{2}^{-} \end{bmatrix} \\ \hline s \cdot (s^{3} + 10 \cdot s^{2} + 49 \cdot s + 10 \\ \hline s + 10 \cdot (s^{2} + 12 \cdot s + 10) \cdot (s^{2} + 10) \cdot$$

4. getNum() and getDenom() (screen 19) give the numerator and denominator, respectively.

CATALOG getNum(h) STO> num

CATALOG getDenom(h) STO> denom

The TI-89 automatically expands these terms, so **cFactor()** must be used again if you want to see the factors.

5. Once the numerator and denominator are separated, the zeros and poles are found by using the **cZeros()** command (screens 20 and 21).

CATALOG cZeros(num , s) STO► zero

CATALOG cZeros(denom , s) STO► pole

Plotting the poles and zeros takes a few steps.

- a. Store the two lists in a data object called pz (screen 22).
- b. pz can't be displayed in the Home screen, but it can be edited by pressing APPS 6:Data/Matrix Editor 2:Open (screen 23).
- c. The first column lists the poles; the second column lists the zeros. To help remember this, add labels to each of the columns by pressing ⊙ ⊙ and typing poles ENTER followed by () ⊙ and typing zeros ENTER (screen 24).

Note getE 1:ge	e: To enter getNum() and Denom() press F2 B:Extract, then tNum(or 2:getDenom(.						
	$ \begin{bmatrix} F_{4}^{+}, F_{2}^{+}, F_{3}^{+}, F_{4}^{+}, F_{5}^{+}, F_{4}^{+}, F_{5}^{+}, F_{4}^{+}, F_{5}^{+}, F_{4}^{+}, F_{5}^{+}, F_{5}$						
(19)	s·(s ⁻ + 10·s ⁻ + 49·s + 100) getDenom(h)⇒denom LAFLACE RAD AUTO FUNC 15/30						
	$ \begin{array}{c} \hline \mathbf{F1}^{+} & \mathbf{F2}^{+} & \mathbf{F3}^{+} & \mathbf{F4}^{+} & \mathbf{F5}^{+} & \mathbf{F6}^{+} \\ \hline \mathbf{F0} & \mathbf{F1} & \mathbf{F1} & \mathbf{F3}^{+} & \mathbf{F1} & \mathbf{F5}^{+} & \mathbf{F6}^{+} \\ \hline \mathbf{getUperom}(\mathbf{n}) \neq \text{denom} \\ \mathbf{s} \cdot \left(\mathbf{s}^{-3} + 10 \cdot \mathbf{s}^{-2} + 49 \cdot \mathbf{s} + 100\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{s}^{-3} + 10 \cdot \mathbf{s}^{-2} + 49 \cdot \mathbf{s} + 100\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} \cdot \mathbf{s}^{-2} + \mathbf{s}^{-2} + 49 \cdot \mathbf{s}^{-2} + 1 - 3\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} \cdot \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} \cdot \mathbf{s}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-4} - \mathbf{s}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2} + \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{c}^{-2} - \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf{s}^{-2} - \mathbf{s}^{-2}\right) \\ \hline \mathbf{s} \cdot \left(\mathbf$						
(20)	cZeros(denom,s)⇒pole LAPLACE RADAUTO FUNC 17/30						
(21)	(-4 0 -3 - 4 ⋅i -3 + 4 ⋅i) cZeros(denom,s)→pole LAPLACE RAD AUTO FUNC 1/17						
$ \begin{array}{c} \hline F_{1}^{2} & F_{2}^{2} & F_{3}^{2} & F_{4}^{2} & F_{5}^{5} & F_{6}^{6} \\ \hline Tools[misebra]ColeBther[PrimtB]Ctean up \\ \hline & cZeros(num, s) \neq zero \\ \hline & + 2 \cdot i & -1 - 3 \cdot i & -1 + 3 \cdot i) \\ \hline & cZeros(denom, s) \neq pole \\ \hline & e^{-4} & 0 & -3 - 4 \cdot i & -3 + 4 \cdot i) \\ \hline & NewData & pz, pole, zero \\ \hline \end{array} $							
(22)	Done NewData pz,pole,zero						
(22) LAPLACE RAD AUTO FUNC 18/30							
(23)	Tooris prot Setup (cin medder (cin cluti) stat Data c1 c2 c3 -6-2*i 3 -3-4*i 1-3+4*i -1-3*i 4 -3+4*i -1+3*i -1-3+i						
	Tobis Pooles Zerros Data poles zeros C1 C2 C3 1 -4 -6-2*i 2 0 -6+2*i 3 -3-4*i -1-3*i 4 -3+4*i -1+3*i						
(24)	C2= LAPLACE RAD AUTO FUNC						

The real part of each pole (or zero) provides the x-component and the imaginary part, the y-component in the complex plane.

- **d.** To separate the poles into their real and imaginary parts, first press () and type **real(c1)** [ENTER]. This makes column **c3** the real part of column **c1**.
- e. Then press () ⊙ imag(c1) ENTER to make column c4 the imaginary part of c1 (screen 25).
- f. Repeat this process for the zeros making column c5 the real part of c2 () → real(c2) ENTER) and column c6 the imaginary part of c2 () → imag(c2) ENTER). Note that the screen scrolls to reveal c5 and c6 (screen 26).
- **g.** To plot the data, press F2 F1 and fill in the required data as shown in screen 27. Press ENTER.

This will plot the real part of the poles (**c3**) versus the imaginary part of the poles (**c4**) as a cross.

- h. Press ⊙ [F1] to set Plot 2 to plot the zeros with boxes (screen 28). Press [ENTER].
- i. Press [WINDOW] to set the plot ranges (screen 29). Turn OFF Grid and Labels with ● 1. Turn off the previous graph with ● [].
- j. Finally, press [GRAPH] to see the poles and zeros graphed in the complex plane (screen 30). This representation is usually called the pole/zero constellation.

	F1+ F2 F3 F4 F5 F6+F7 ToolsPlot SetupCellHeaderCalcUtilStat					
	DATA	poles	zeros			
		c1	c2	c3	c4	
	1	-4	-6-2	-4	0	
	2	0	-6+2	0	0	
	3	-3-4	-1-3	-3	-4	
	4	-3+4	-1+3	-3	4	
(25)	<u>r1c1=-4</u>					
(20) LAPLACE RAD AUTO FUNC						

Note: Press • [] and select a cell width of 5 to see four columns.



Topic 30: Frequency Response

The frequency response that corresponds to the pole/zero constellation in Topic 29 is graphed by noting that

$$|H(j\omega)| = |H(s)|_{s=j\omega}$$

1. To do this, press HOME and enter the equation as shown in screen 31.

[CATALOG] abs(h]) [] s = 2nd [*i*] w STO► eqn

- Enter eqn as the function y1(x) (screen 32).
 eqn [] w = x (STO►) y1 ((x))
- **3.** Press [Y=] to verify this. Be sure to deselect plots 1 and 2 in the Y= Editor using F4 (screen 33).
- **4.** Press [WINDOW] to set the correct graphing parameters in the Window Editor (screen 34).
- **5.** Press [GRAPH] to display the graph of frequency response (screen 35).

Notice that the effects of the pole farthest from the axis can be seen as slight rises near the left and right sides of the graph. The zeros are causing the dips around $x=\pm 3$, and the pole at the origin is causing the large peak in the middle.



Topic 31: 3D Poles and Zeros

A different perspective of H(s) is gained from a 3D graph where the z-axis represents the magnitude of H(s).

- 1. To do so, press MODE () and select 5:3D (screen 36). Press ENTER.
- 2. Press [Y=] and enter the function to be graphed (screen 37).

CATALOG abs($h \mid s = x + 2nd [i] \times y$)

- 3. Press [WINDOW] and set the x, y, and z scales (screen 38). Note that these are the default values, except zmin has been set to 0.
- Finally, press [GRAPH] (screen 39). It will take a few minutes for the graph to display. Once the graph is complete, press] and select AXES and turn ON the Labels.
- 5. The three poles are clearly visible. Things to try:
 Press X, Y, or Z to look down the corresponding axis.
 Use the cursor controls (④ ④ ⊙ ⊙) to spin the graph.
 Press ① to return to the original view.
- 6. Press 1 and change the Style to HIDDEN SURFACE (screens 40 and 41).



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- Press [] and change the Style to WIRE AND CONTOUR to see contours highlighted on the graph (screen 42). This will take a few minutes to recalculate.
- 8. Press (2) while in WIRE AND CONTOUR mode to view the contours from above (screen 43). Press (1) to return to the original view.
- **9.** Press [WINDOW], set **xgrid** and **ygrid** to larger values (25 in this case), and press [GRAPH] to get a smoother graph (screen 44). This also takes a few minutes to recalculate.
- To zoom in (screen 45), press the x key (the multiplication key, not the letter x).



Tips and Generalizations

The TI-89's symbolic math capability makes it a good choice for manipulating equations in the s-domain. The key step to plotting on the s-plane (real vs. imaginary) is to use the "with" operator ([]) to replace **s** with $\mathbf{x} + [\iota] \mathbf{y}$. Although plotting |H(s)| is most common, the TI-89 can just as easily plot the angle of H(s) by entering

angle(h (s) \square s = x + [i] y).

Although these examples solved for a single node problem with only one equation, v(s), more complex circuits with more nodes (and therefore more equations) also can be solved.

The TI-89 assisted the conversion from the s-domain to the time-domain by doing the partial fraction expansion. Chapter 7 shows how to find a system's response by staying in the time-domain and using convolution.